

Current Source Inverter-Fed Variable Reluctance Motor Drive with DC-Machine-Like Control Characteristics

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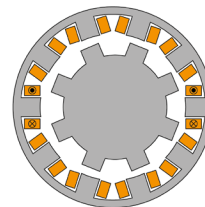
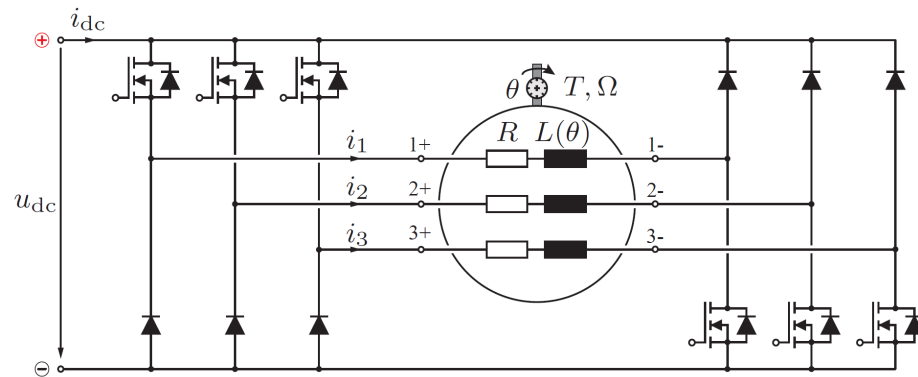
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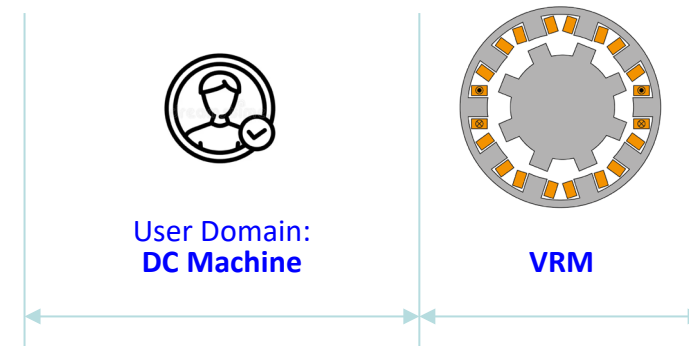
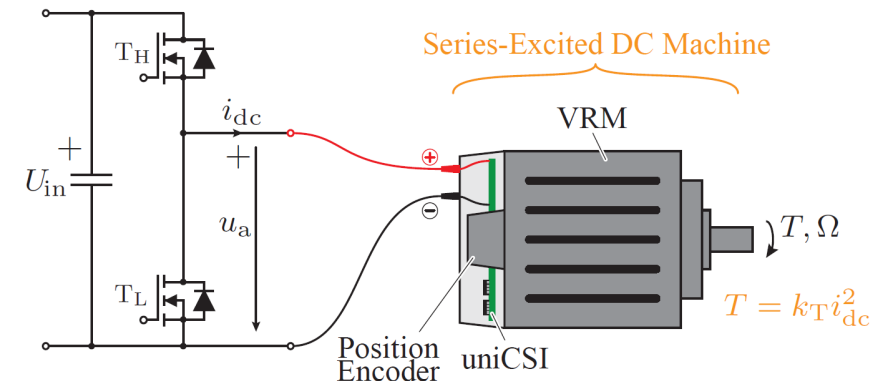
Motivation: Make Reluctance Motor Behave Like a DC Machine for the User

- Variable reluctance machine (VRM) operation → challenge with controlling the currents especially at high speeds
- Non standard inverter topology due to unipolar currents → there are many inverter topology options for VRMs



VRM
(Variable
Reluctance
Machine)

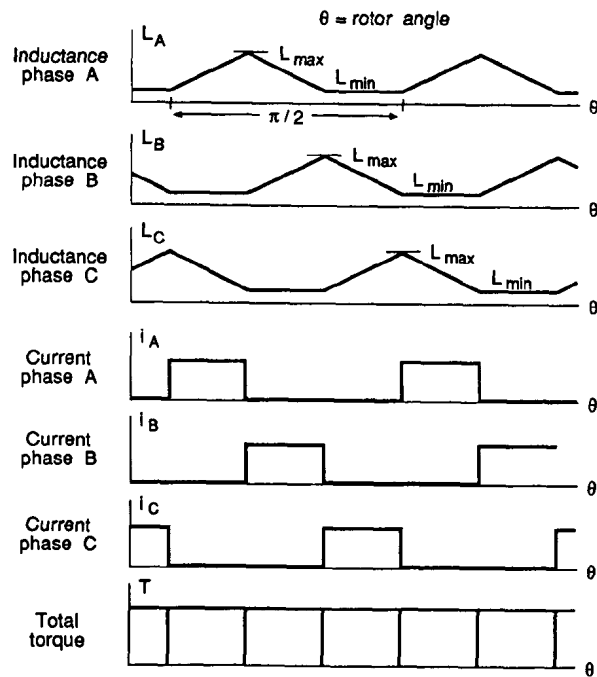
- Our goal is to make VRM simple as DC machine for the user →



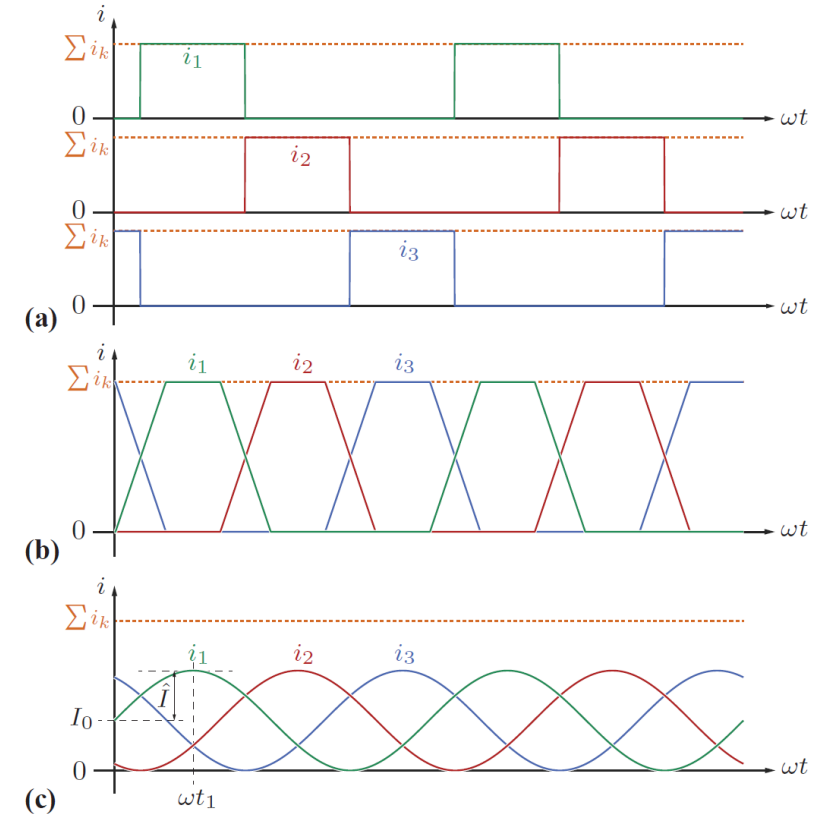
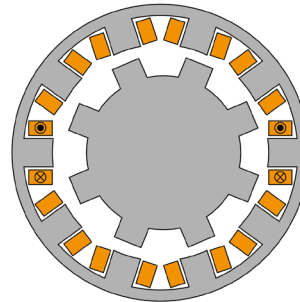


Reluctance Machine: Unipolar Currents

- Built from iron and copper → no permanent magnets; good ratio of the torque and the moment of inertia → good acceleration
- Torque is proportional to the inductance change and the square of the current → unipolar phase currents



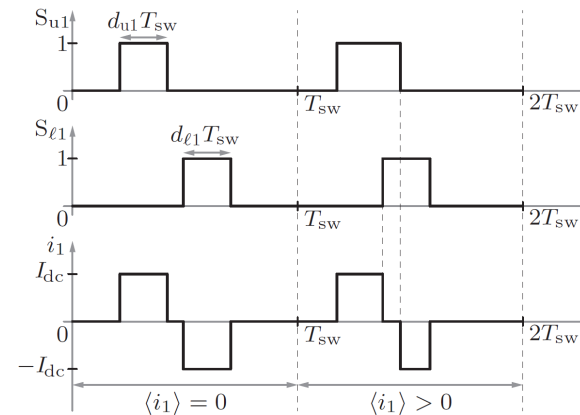
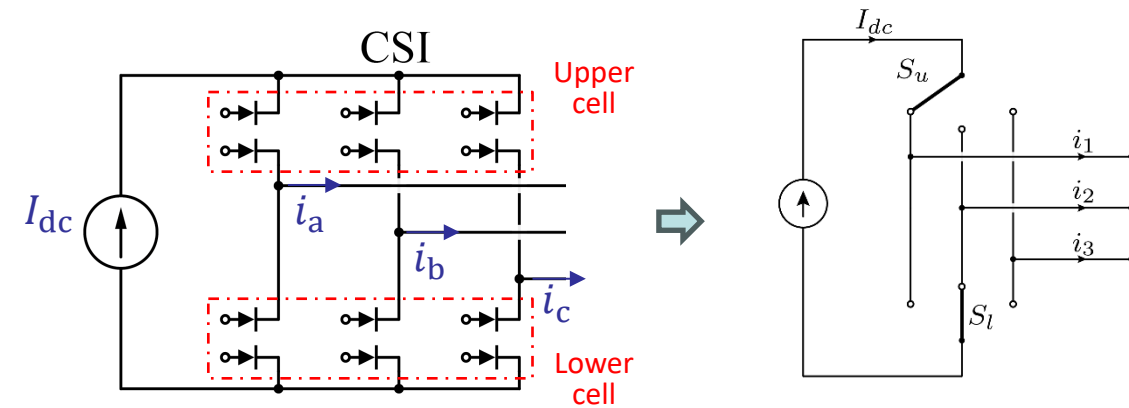
$$T \sim i^2 \frac{\partial L}{\partial \theta}$$





Current Source Inverter (CSI) Modulation

- The task of the modulation is to ensure the desired value of the average phase current at the output
- Upper switch on → positive current pulse; lower switch on → negative current pulse; both switches on → zero current
- Every switching state must ensure the 'flow' of the DC link current



Average value of the current $\langle i_1 \rangle$ is obtained by averaging the DC link current pulses.

$$\langle i_1 \rangle = I_{dc}(d_{u1} - d_{l1})$$

- Current source inverter modulation process has to ensure the following two conditions:

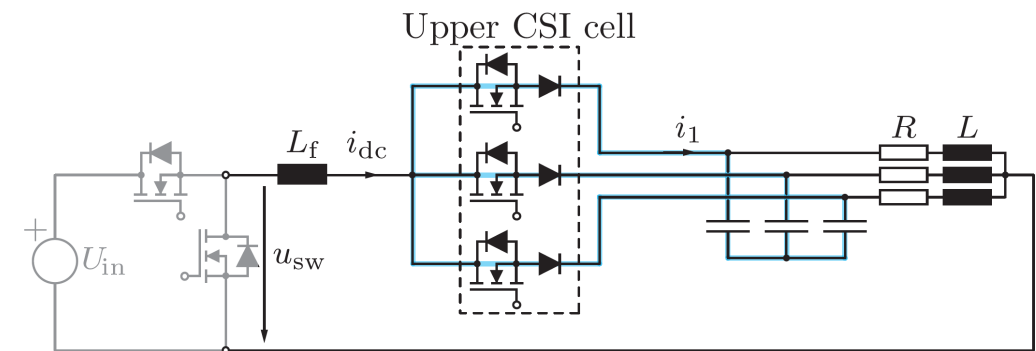
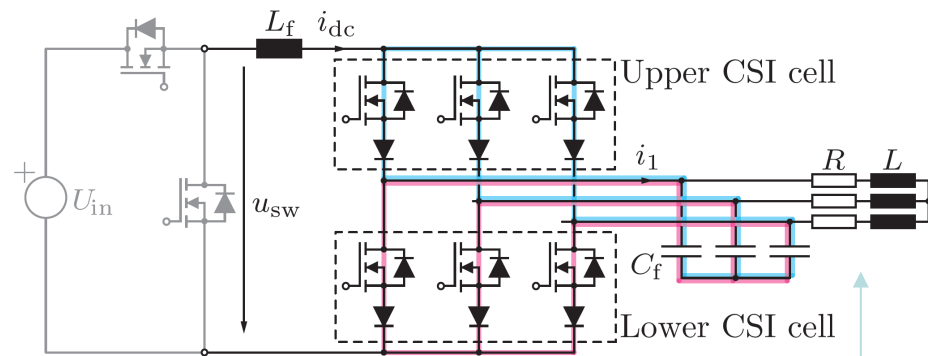
1. Average value of the phase current: $\langle i_1 \rangle = I_{dc}(d_{u1} - d_{l1})$
2. Continuity of the DC link current: $d_{u1} + d_{u2} + d_{u3} = 1$ and $d_{l1} + d_{l2} + d_{l3} = 1$



uniCSI: Current Source Inverter for Unipolar Currents (1)

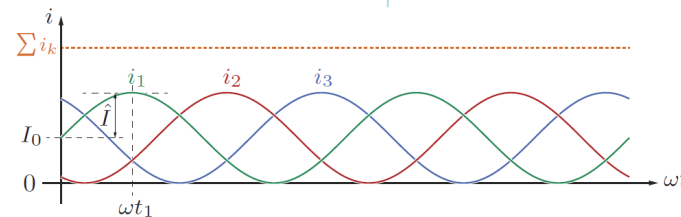
- The average phase current is proportional to the duty cycle difference of the upper and lower CSI cells
- CSI needs a DC link inductor (L_f) so we can build a DC link current source!
- If we need positive only current → we can realize it with the upper cell only!
- The RL load (represents VRM) need positive only phase currents!

$$\langle i_1 \rangle = i_{dc}(d_{u1} - d_{l1})$$



Unipolar positive
currents

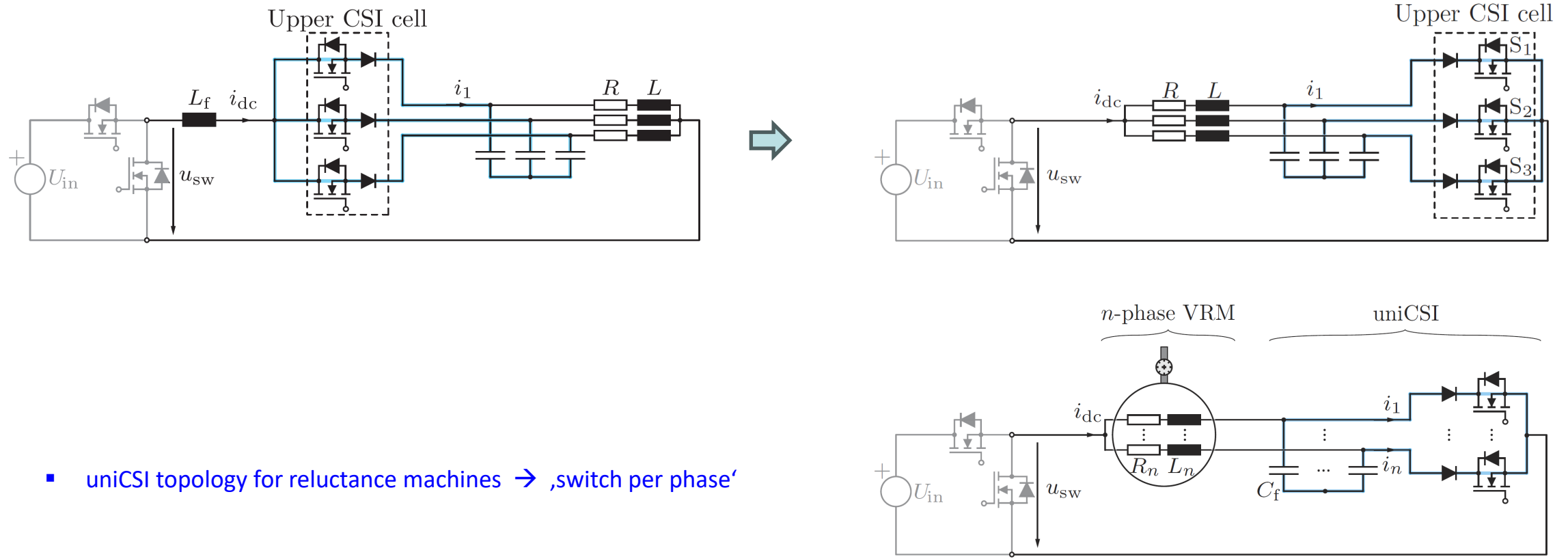
How to further improve this topology? →





uniCSI: Current Source Inverter for Unipolar Currents (2)

- Reluctance machine has open end windings due to unipolar currents → remove the DC link inductor as machine common-mode inductance acts as L_f
- Series connection of the upper CSI cell and the RL load → relocate them so that MOSFETs are connected to the GND (no gate driver isolation needed)

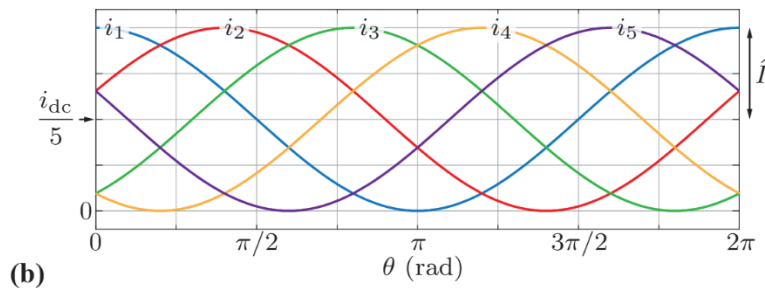
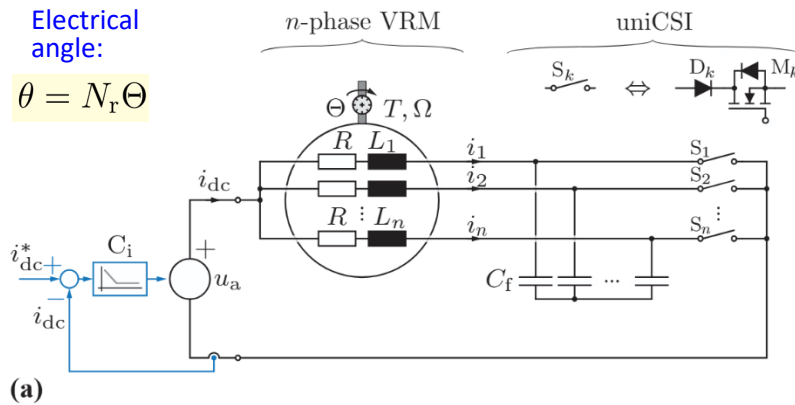


- uniCSI topology for reluctance machines → 'switch per phase'



Variable Reluctance Machine (VRM) Drive with uniCSI

- Switch per phase uniCSI topology to supply VRM → switch is series connection of a diode and a MOSFET (or bidirectional switch)
- For VRM, the phase currents sum up to the DC link current value → for 5-phase example: $i_1 + i_2 + i_3 + i_4 + i_5 = i_{dc}$
- We assume sinusoidal currents, but other waveforms like trapezoidal are possible



$$i_k = d_k i_{dc}$$

$$\sum_{k=1}^5 i_k = i_{dc}$$

Phase currents:

$$i_k = m \frac{i_{dc}}{n} \cos \left(\theta + \theta_i - (k-1) \frac{2\pi}{n} \right) + \frac{i_{dc}}{n}$$

uniCSI duty cycles:

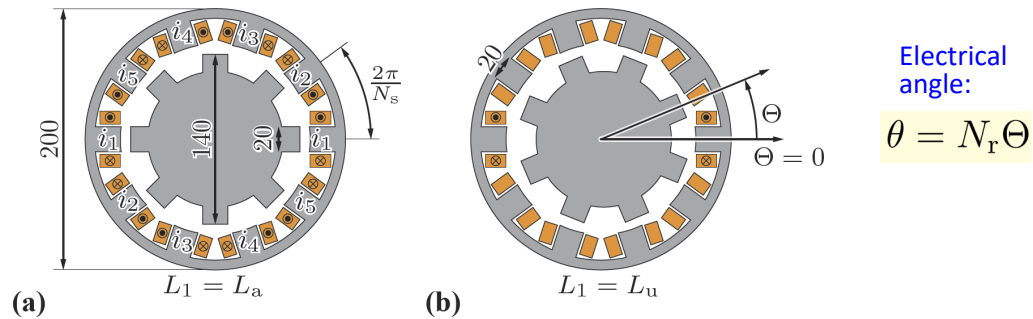
$$d_k = \frac{i_k}{i_{dc}} = \frac{m}{n} \cos \left(\theta + \theta_i - (k-1) \frac{2\pi}{n} \right) + \frac{1}{n}$$

$$\sum_{k=1}^n d_k = 1$$



Variable Reluctance Machine Model

- We assume only **self-inductance** to be contributing to the **flux linkage**, which is justified by FEM simulations where the flux due to mutual inductances is negligible.
- In this example $N_r = 8$

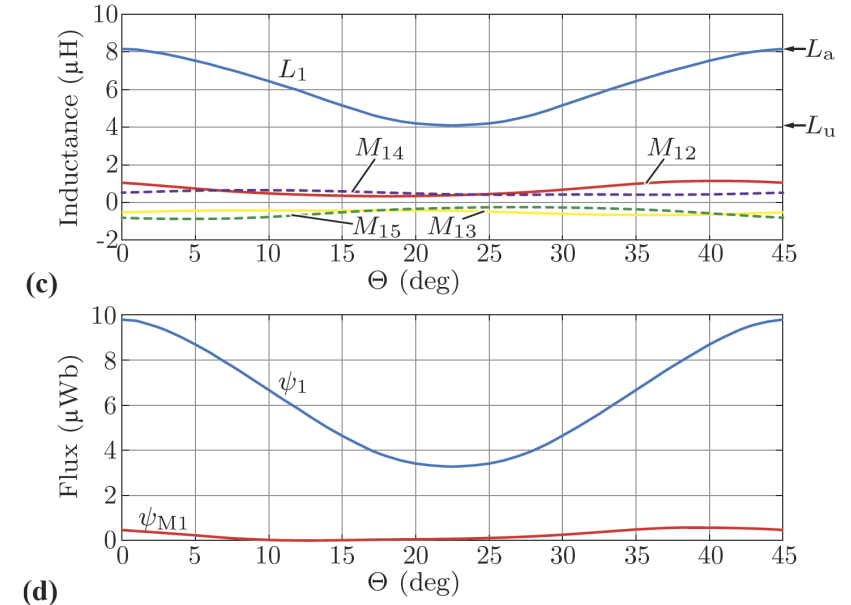


- Sinusoidal inductance (used also in PLECS):

$$L_k(\theta) = L_u + (L_a - L_u) f_k(\theta)$$



$$f_k(\theta) = \frac{1}{2} + \frac{1}{2} \cos \left(\theta + (k-1) \frac{2\pi}{n} \right)$$



Flux linkage: $\psi_k(\theta) = L_k(\theta) i_k(\theta)$

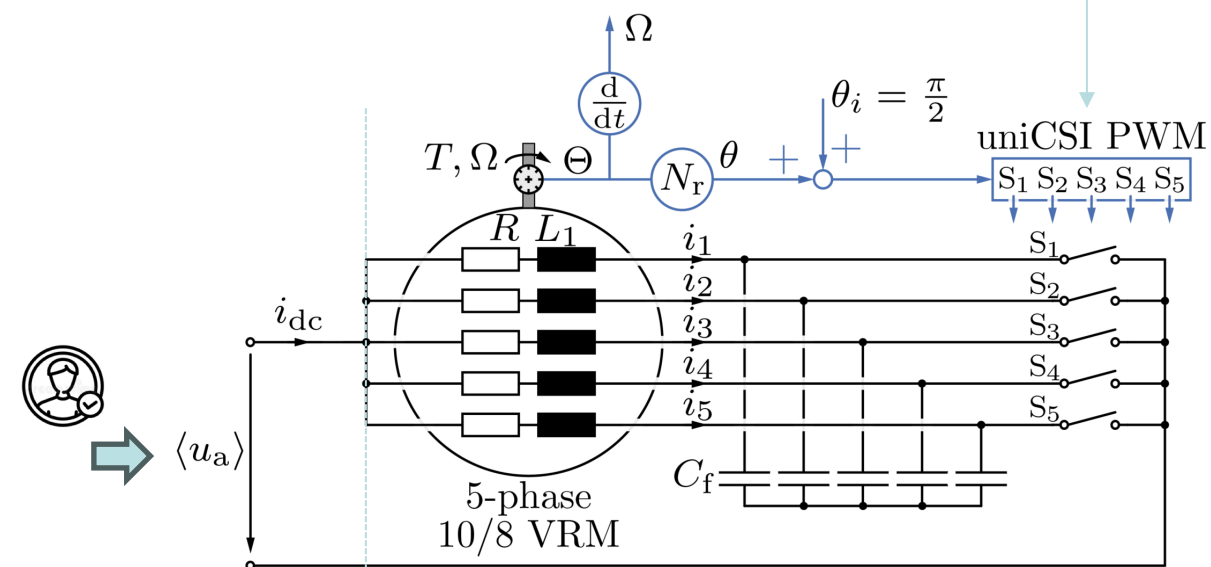
Voltage equation used for modeling: $u_k = R i_k + \frac{d\psi_k}{dt}$



DC-Side of the uniCSI

- Electrical angle given by the rotor position as: $\theta = N_r \Theta$

$$d_k = \frac{i_k}{i_{dc}} = \frac{m}{n} \cos \left(\theta + \theta_i - (k-1) \frac{2\pi}{n} \right) + \frac{1}{n}$$



Power balance:

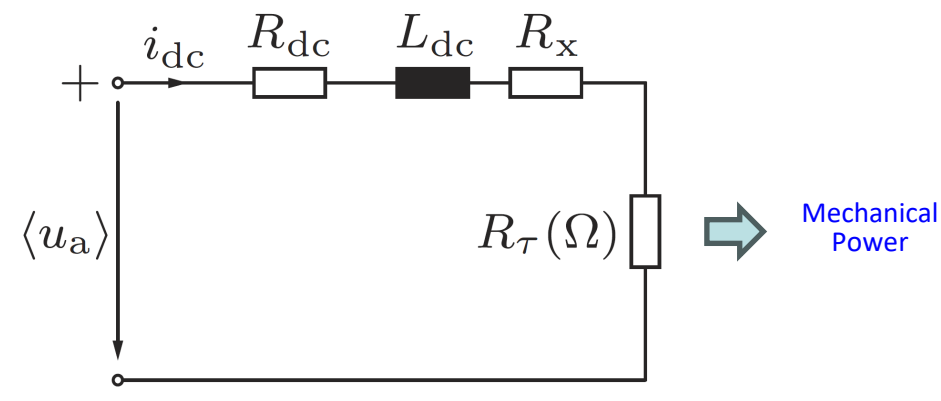
$$\langle u_a \rangle i_{dc} = \sum_{k=1}^5 u_k i_k$$

$$u_k = R i_k + \frac{d\psi_k}{dt}$$

$$\langle u_a \rangle i_{dc} = R \sum_{k=1}^5 i_k^2 = \underbrace{\frac{m^2 + 2}{10}}_{=R_{dc}} R i_{dc}^2$$

$$\begin{aligned} \langle u_a \rangle i_{dc} &= \sum_{k=1}^5 \frac{d\psi_k}{dt} i_k \\ &= \left(L_{dc} \frac{di_{dc}}{dt} + R_x i_{dc} + R_\tau i_{dc} \right) i_{dc} \end{aligned}$$

- VRMs DC-side equivalent circuit:





Torque of the VRM Supplied by uniCSI

- Equivalent circuit, expressions of circuit elements:
- m – uniCSI modulation index
- θ_i – current phase shift with respect to the shaft electrical angle θ

$$L_{\Sigma} = L_a + L_u \quad L_{\Delta} = L_a - L_u$$

Equivalent DC
inductance:

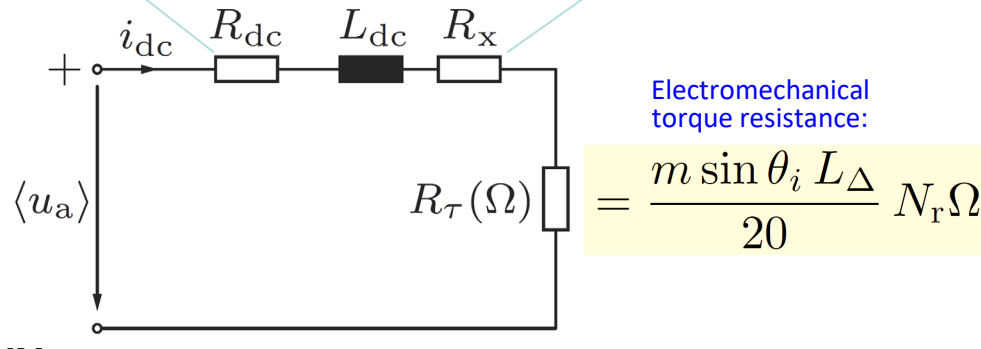
$$L_{dc} = \frac{(m^2 + 2)L_{\Sigma} + 2m \cos \theta_i L_{\Delta}}{20}$$

Dynamic
resistance:

$$R_x = \frac{m L_{\Sigma} + \cos \theta_i L_{\Delta}}{20} \frac{dm}{dt}$$

Equivalent DC
resistance:

$$R_{dc} = \frac{(m^2 + 2)}{10} R$$



Electromechanical
torque resistance:

$$= \frac{m \sin \theta_i L_{\Delta}}{20} N_r \Omega$$

- Torque expression derivation → power balance

$$R_{\tau} i_{dc}^2 = T \Omega = T \frac{\omega}{N_r}$$



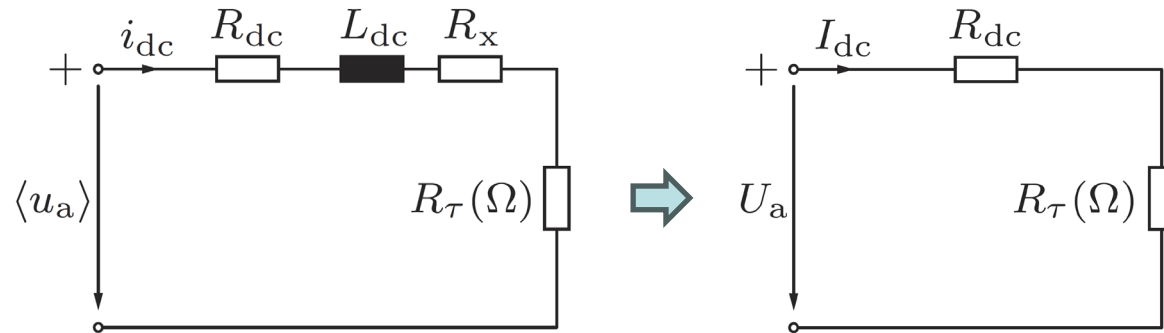
$$T = \frac{2}{5} m \sin \theta_i L_{\Delta} i_{dc}^2$$

$$k_T = \frac{2}{5} m \sin \theta_i L_{\Delta}$$



DC-Side Speed-Torque Characteristic

- Speed-torque characteristics are derived for steady state → inductor (L_{dc}) and dynamic resistance (R_x) disappear from the steady-state equiv. circuit



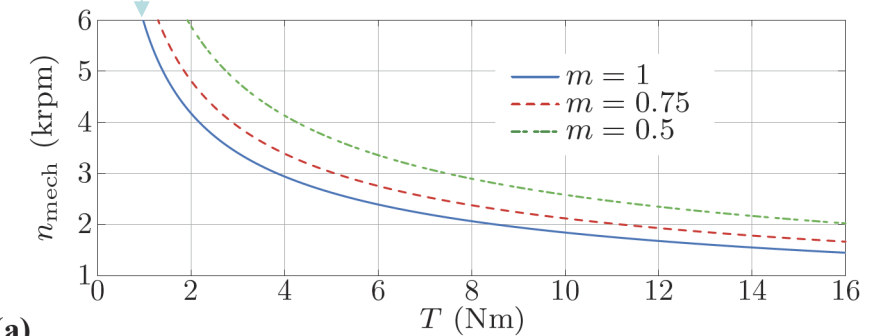
$$U_a = R_{dc} I_{dc} + R_\tau(\Omega) I_{dc}$$

$$R_\tau(\Omega) = k_T \Omega$$

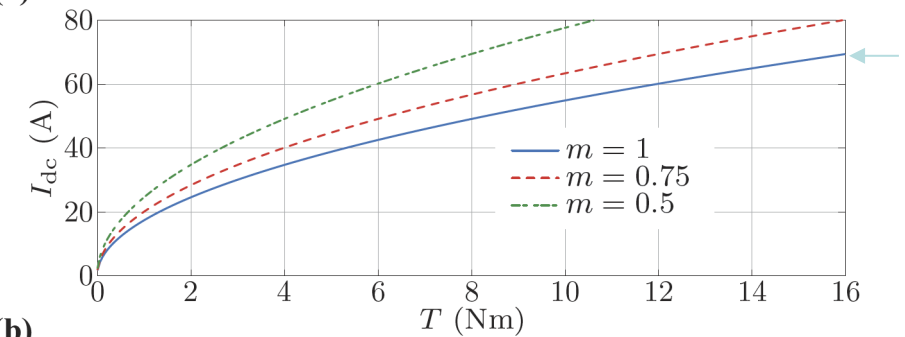
$$I_{dc} = \sqrt{T/k_T}$$

$$\Omega(T) = \frac{U_a}{\sqrt{k_T T}} - \frac{R_{dc}}{k_T}$$

DC-side speed-torque characteristics



(a)



(b)

Best: $m = 1$



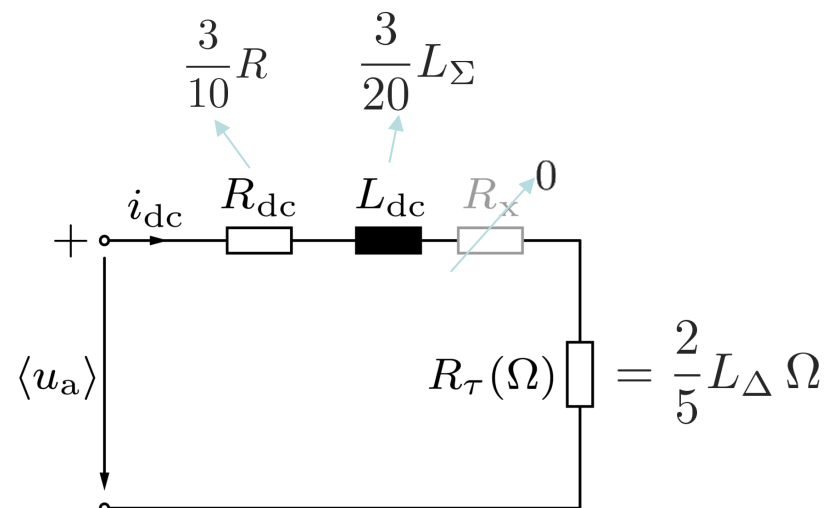
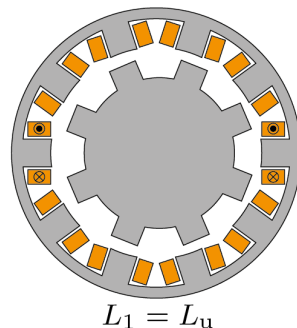
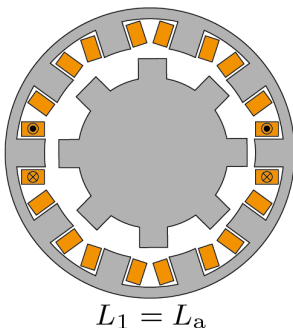
Maximizing Torque per Current

- We choose m and θ_i so that we get maximum torque per ampere \rightarrow we maximise the torque constant

$$T = \underbrace{\frac{2}{5} m \sin \theta_i L_{\Delta} i_{dc}^2}_{\frac{2}{5} L_{\Delta}} \Rightarrow m = M = 1 \quad \theta_i = \theta_I = \frac{\pi}{2}$$

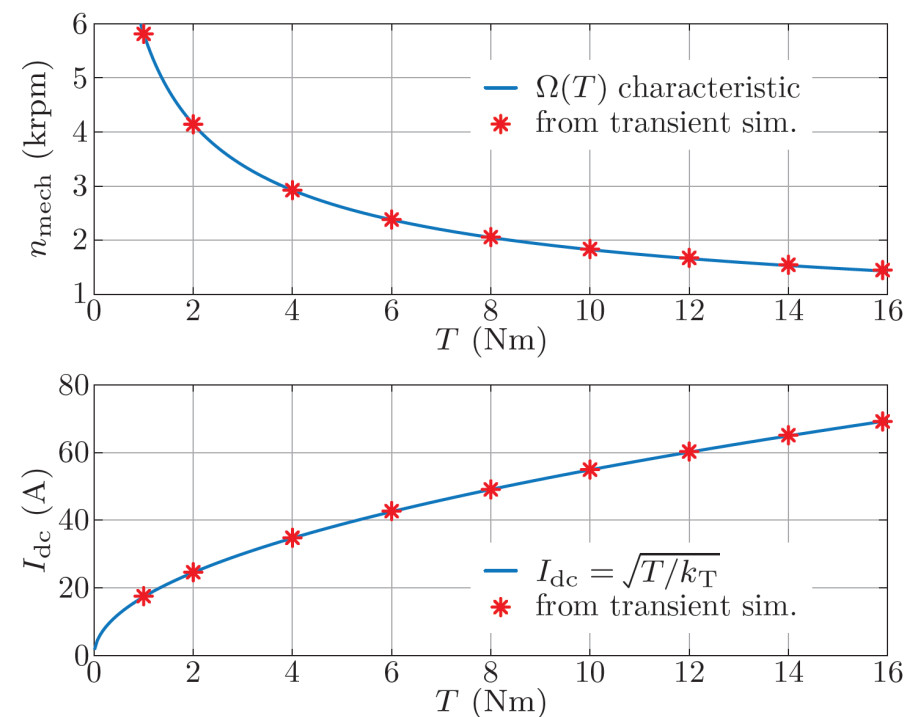
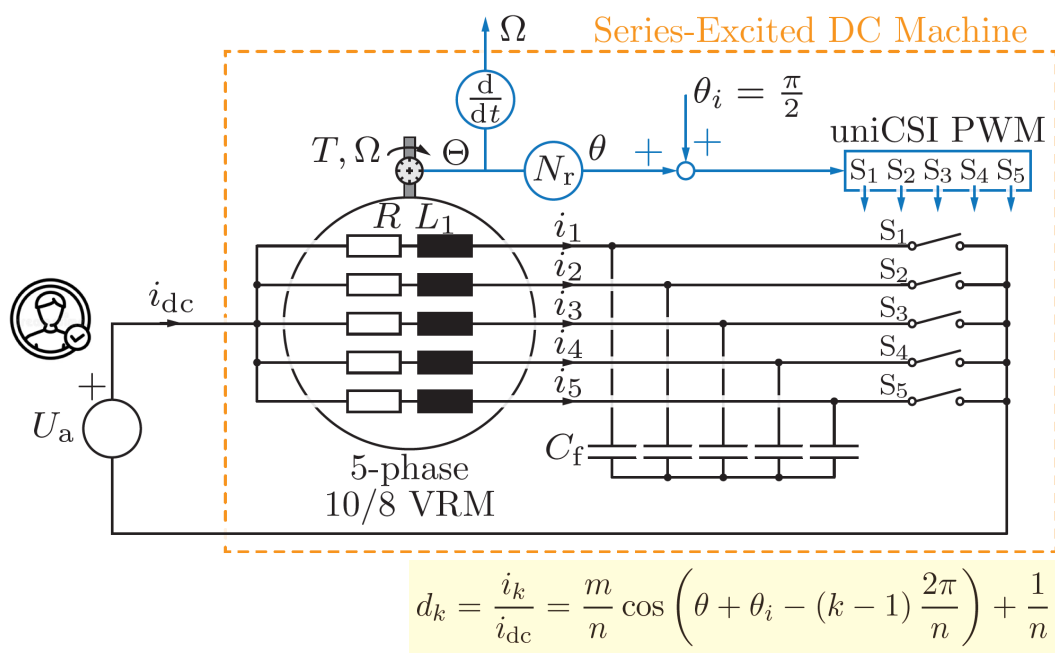
Difference between the aligned
and unaligned inductances

$$L_{\Delta} = L_a - L_u$$



Open-Loop Operation of the uniCSI Supplied VRM

- We fixed modulation index $m = 1$ and the phase shift $\theta_i = \pi/2 \rightarrow$ phase currents are open-loop locked to the motor's shaft position
- This results that the uniCSI supplied VRM behaves like a series-excited DC machine from the DC side connections
- This enables a user to operate such drive system from the DC connections \rightarrow no user interaction with the VRM and uniCSI

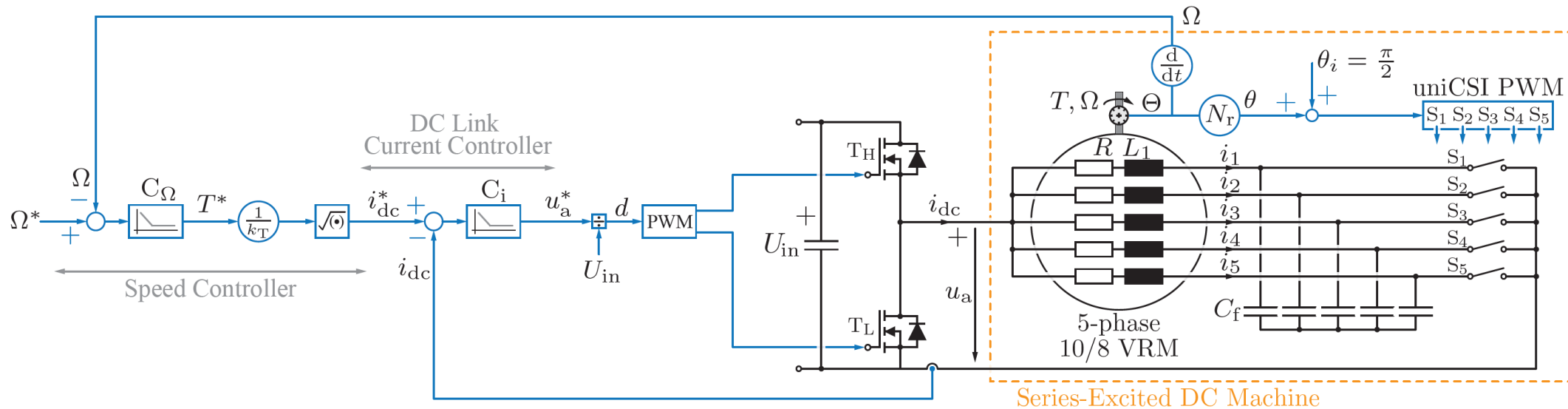


Each star represents a steady state value of the transient time domain simulation!



Speed Control with uniCSI-VRM Drive

- Torque and therefore speed control is managed with the input converter, like with a conventional DC machine
- Open-loop operation of the uniCSI → no controllers | constant $m = 1$ | constant $\theta_i = \pi/2$ | no user interaction with uniCSI
- This drive systems decouples any expertise of reluctance machines from the user → 'plug-and-play' DC drive system



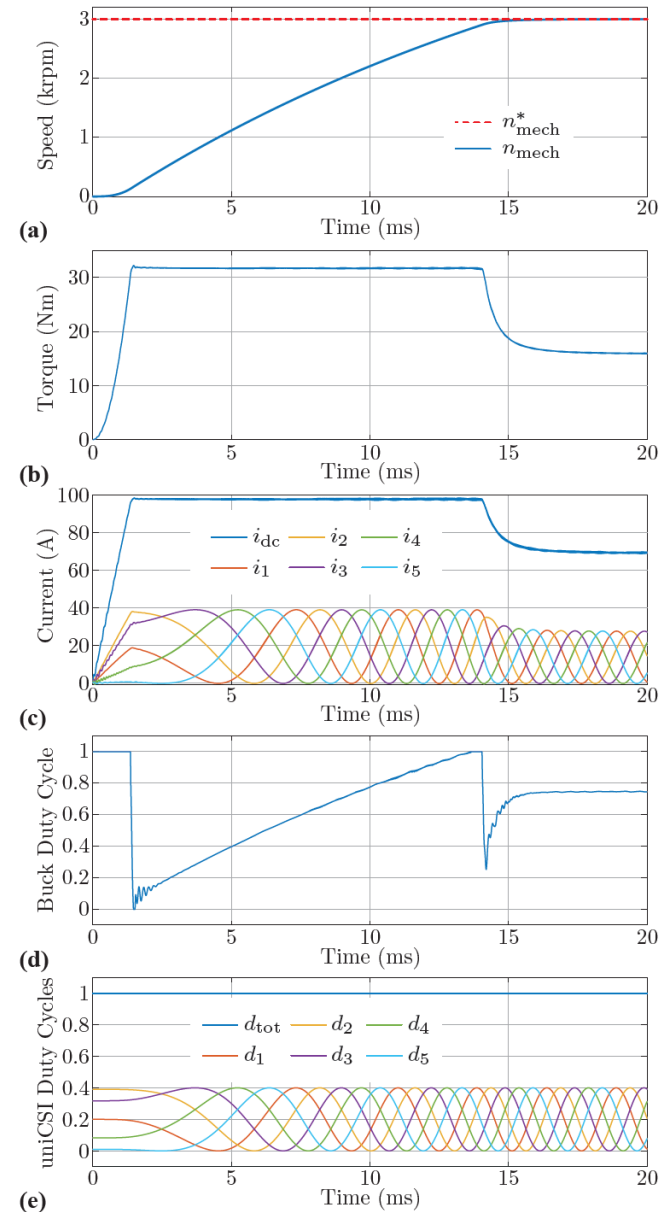


Simulation Results – Speed Transient

- Speed transient from standstill to nominal speed
- Quadratic relationship between the DC link current and the torque

Parameter	Symbol	Value
Buck		
Input voltage	U_{in}	100 V
Switching frequency	$f_{sw,b}$	300 kHz
CSI		
Output capacitance	C_f	0.2 μ F
Switching frequency	f_{sw}	300 kHz
Modulation index	M	1
Current angle	θ_I	$\pi/2$
VRM		
Phase resistance	R	0.05 Ω
Unaligned inductance	L_u	0.5 mH
Aligned inductance	L_a	8.8 mH
Number of stator teeth	N_s	10
Number of rotor teeth	N_r	8
Moment of inertia	J	0.001 kgm ²
Nominal mech. power	P_{mech}	5 kW
Nominal mech. speed	n_{mech}	3000 rpm
DC-side		
DC-side resistance	R_{dc}	0.015 Ω
DC-side inductance	L_{dc}	1.395 mH
Torque constant	k_T	3.32 mN m A ⁻²
Controller gains		
C_i closed-loop bandwidth	f_{cc}	5 kHz
C_i proportional gain	K_{pc}	43.82 V/A
C_i integral gain	K_{ic}	33 143 V/(As)
C_Ω cross-over frequency	f_{cs}	0.5 kHz
C_Ω proportional gain	K_{ps}	3.14 s N m
C_Ω integral gain	K_{is}	1974 N m

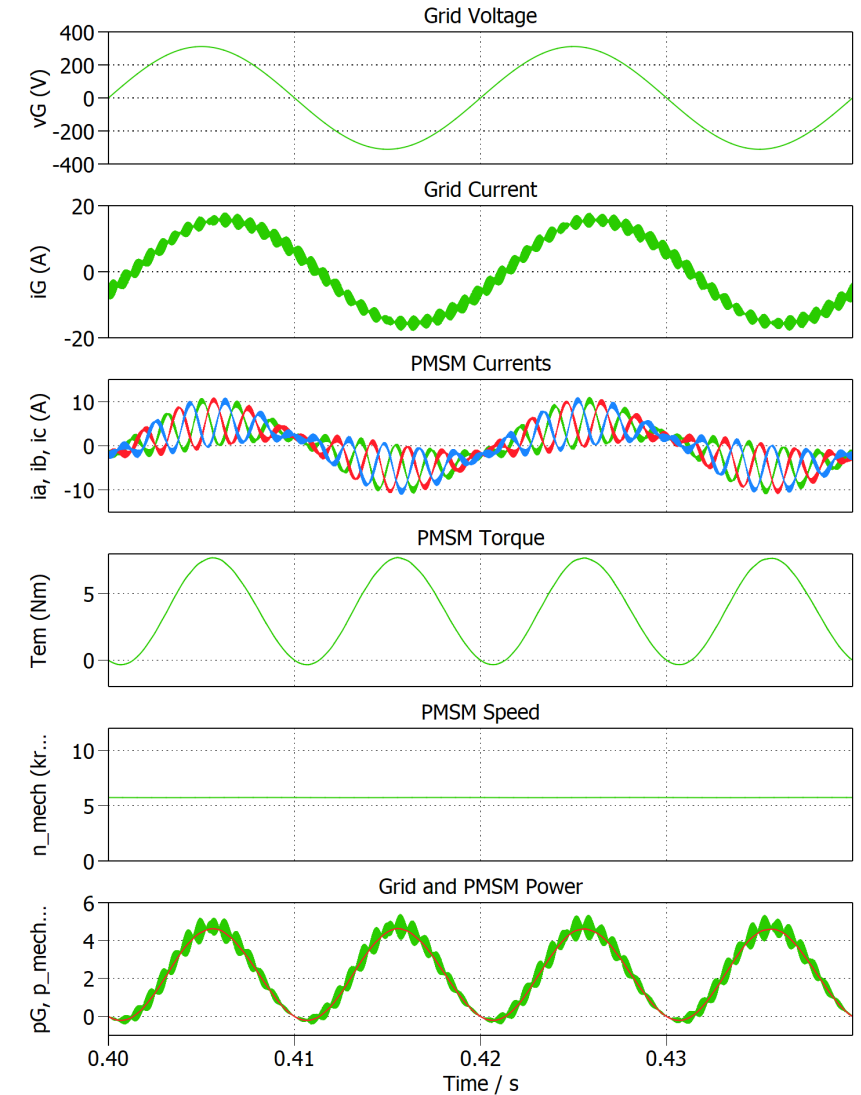
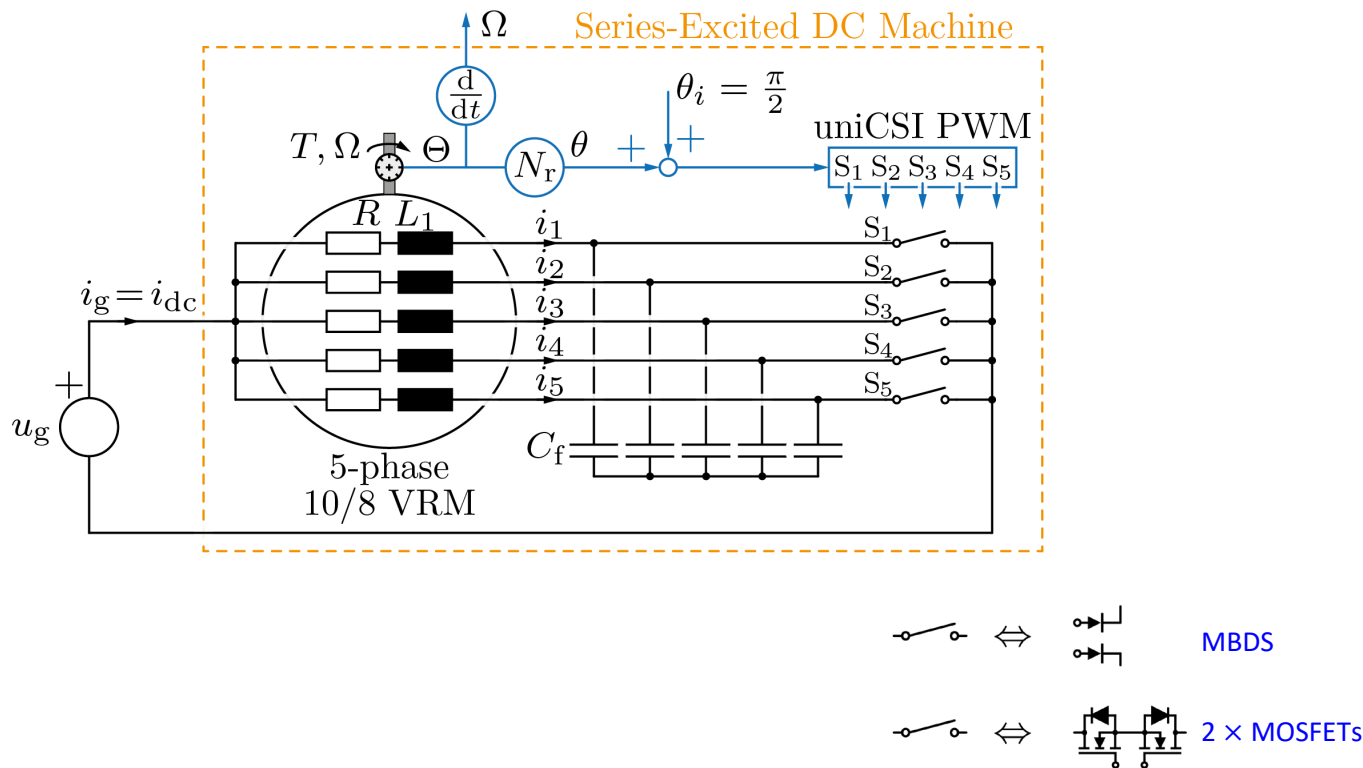
TABLE II: Simulation Parameters.





Grid Supplied uniCSI-VRM: Universal Machine

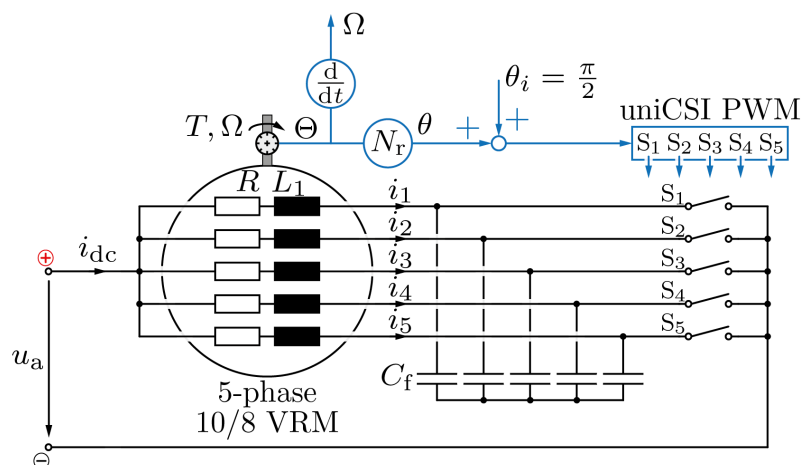
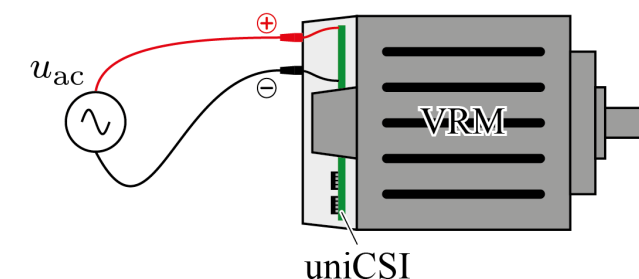
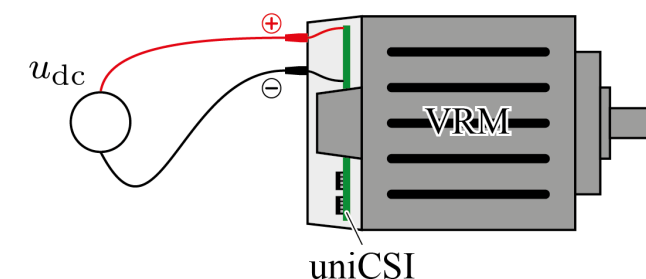
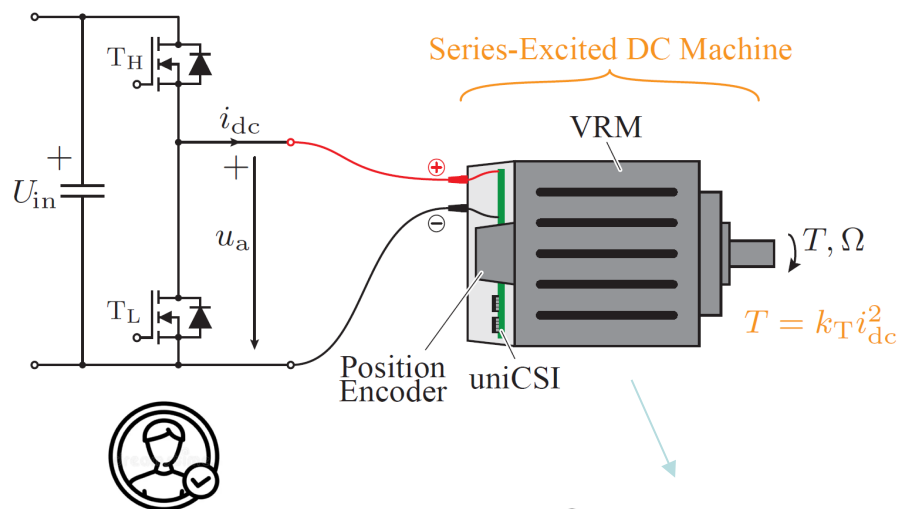
- Open loop controlled uniCSI behaves like a series excited DC machine
- Series excited DC machine → universal machine
- Grid power pulsation → buffered in drive train's inertia
- Future work → speed regulation capabilities





DC Machine Behavior of the uniCSI VRM

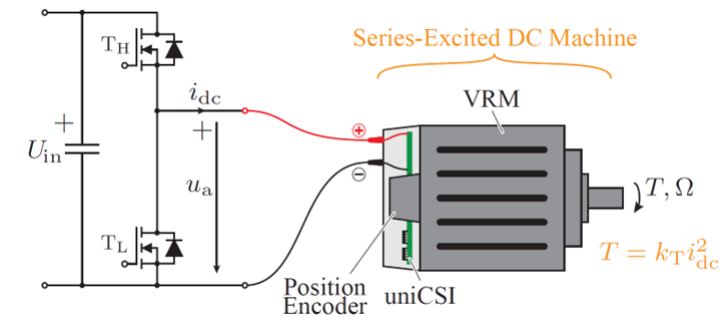
- With shown open loop control of the uniCSI, the whole drive behaves like a series excited DC machine (universal motor)
- High end speed control applications → DC link current control
- Simple applications → it can run without the DC link current control from DC and AC grid (ongoing work)



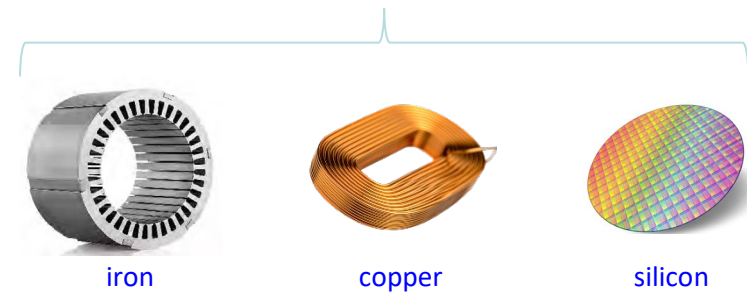


Conclusions

- Novel uniCSI topology
- Open-loop control
- DC-side speed torque characteristic
- No user interaction with the inverter nor VRM control
- User 'sees' a DC machine
- Speed-controlled drive system through input converter
- Universal machine
- Future work: Hardware verification and DC/AC voltage supply operation



Future motor building materials:



Thank you!

